

Data Sheet March 3, 2009 FN6732.0

Digital Proximity Sensor with Interrupt Function

The ISL29021 is an integrated proximity and infrared sensor with a built-in IR LED driver and I²C Interface (SMBus Compatible). This device provides infrared sensing to allow proximity estimation featured with interrupt function.

For infrared and proximity sensing, an internal ADC has been designed based on the charge-balancing A/D conversion technique.

The ADC is used to digitize the output signal from the photodiode array when the internal IR LED driver is turned on and off for the programmed time periods under user-selected modulation frequency to drive the external IR LED. As this proximity sensor employs a noise cancellation scheme to highly reject unwanted IR noise, the digital output of proximity sensing decreases with distance. The driver output current is user selectable up to 100mA to drive different types of IR emitters LEDs.

Four different modes of operation can be selected via the I²C interface: programmable IR sensing once, programmable proximity sensing once, programmable continuous IR sensing and programmable continuous proximity sensing. The programmable one-time operation modes greatly reduce power because an immediate automatic shutdown reduces overall supply current less than 0.5µA.

The ISL29021 supports both hardware and software interrupts that remain asserted until the host clears it through I²C interface for proximity detection.

Designed to operate on supplies from 2.5V to 3.63V, the ISL29021 is specified for operation over the -40°C to +85°C ambient temperature range. It is packaged in a clear, Pb-free 8 Ld ODFN package.

Ordering Information

| PART NUMBER (Note) | TEMP. RANGE (°C) | PACKAGE (Pb-Free) | PKG. DWG. # | |
|-----------------------|---------------------|----------------------|----------------|--|
| ISL29021IROZ-T7* | -40 to +85 | 8 Ld ODFN | L8.2.1x2.0 | |
| ISL29021IROZ-EVALZ | Evaluation Board | | | |

^{*}Please refer to TB347 for details on reel specifications.

NOTE: These Intersil Pb-free plastic packaged products employ special Pb-free material sets; molding compounds/die attach materials and NiPdAu plate - e4 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations. Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

Features

Proximity Sensing

- Ambient IR Cancellation During Proximity Sensing
 - Works Under Direct Sunlight
- · IR LED Driver with Programmable Source Current
 - Adjustable Current Drive from 100mA to 12.5mA
- Programmable LED current Modulation Frequency
- Variable Conversion Resolution up to 16-bits
- Selectable Range (via I²C)
- · Works Under Various Light Sources, Including Sunlight

Ideal Spectral Response for Proximity Sensor

- Proximity sensor range from 850nm to 950nm
 - Can use either 850nm or 950nm LED solution

Ultra Low Power

- · 90µA Max Operating Current
- · Software Shutdown and Automatic Shutdown
 - 0.5µA Max Shutdown Current

Easy to Use

- I²C (SMBus Compatible) Output
- · No Complex Algorithms Needed
- · Temperature Compensated
- · Small Form Factor
 - 8 Ld 2.0mmx2.1mmx0.7mm ODFN Package

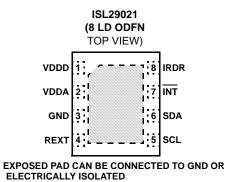
Additional Features

- I²C and SMBus Compatible
- 1.7V to 3.63V Supply for I²C Interface
- · 2.25V to 3.63V Sensor Power Supply
- · Pb-Free (RoHS compliant)

Applications

- · Display and Keypad Proximity Sensing for:
 - Mobile Devices: Smart Phone, PDA, GPS
 - Computing Devices: Notebook PC, Webpad
 - Consumer Devices: LCD-TV, Digital Picture Frame, Digital Camera
- Industrial and Medical Proximity Sensing

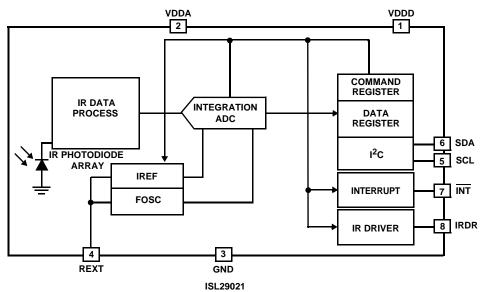
Pinout



Pin Descriptions

| PIN NUMBER | PIN NAME | | DESCRIPTION | | | | |
|------------|----------|--|--|--|--|--|--|
| 1 | VDDD | Positive digital supply: 2.5V to 3 | sitive digital supply: 2.5V to 3.63V. | | | | |
| 2 | VDDA | Positive analog supply: 2.5V to | 3.63V, VDDA and VDDD should be externally shorted. | | | | |
| 3 | GND | Ground. | | | | | |
| 4 | REXT | External resistor pin setting the i is recommended. | ernal resistor pin setting the internal reference current and the conversion time. 499k Ω with 1% tolerance resistor ecommended. | | | | |
| 5 | SCL | I ² C serial clock line | The I ² C bus lines can be pulled from 1.7V to above V _{DD} , 3.63V max. | | | | |
| 6 | SDA | I ² C serial data line | C serial data line | | | | |
| 7 | ĪNT | Interrupt pin; LO for interrupt/ala | nterrupt pin; LO for interrupt/alarming. The $\overline{\text{INT}}$ pin is an open drain. | | | | |
| 8 | IRDR | IR LED driver pin connecting to programmed through I ² C. | R LED driver pin connecting to the anode of the external IR LED. The source current of the IR LED driver can be programmed through I ² C. | | | | |
| | | Exposed pad connected to grou | xposed pad connected to ground or electrically isolated. | | | | |

Block Diagram



Absolute Maximum Ratings (T_A = +25°C)

| $V_{SUP(VDDD,VDDA)}$ Supply Voltage between V_{DD} and GND 4V |
|--|
| V _{DDA} Supply Voltage between V _{DDA} and GND V _{DDD} +/- 0.5V |
| I ² C Bus (SCL, SDA) and INT Pin Voltage0.2V to 4V |
| I ² C Bus (SCL, SDA) and INT Pin Current <10mA |
| IRDR Pin Voltage0.2V to V _{DD} + 0.5V |
| R _{EXT} Pin Voltage0.2V to V _{DD} + 0.5V |
| ESD Rating |
| Human Body Model |

Thermal Information

| Thermal Resistance (Typical, Note 1) | θ _{JA} (°C/W) |
|--|------------------------|
| 8 Ld ODFN | 88 |
| Maximum Die Temperature | +90°C |
| Storage Temperature | °C to +100°C |
| Operating Temperature | 0°C to +85°C |
| Pb-Free Reflow Profilese | ee link below |
| http://www.intersil.com/pbfree/Pb-FreeReflow.asp | |

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTE:

1. θ_{JA} is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See Tech Brief TB379.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

Electrical Specifications

 $V_{SUP(VDDD,VDDA)} = 3V, T_A = +25^{\circ}C, \ R_{EXT} = 499k\Omega \ 1\% \ tolerance, \ 16-bit \ ADC \ operation, \ unless \ otherwise$ specified.

| PARAMETER | DESCRIPTION | CONDITION | MIN | TYP | MAX | UNIT |
|--|---|---|-------|-----------------------|-------|--------|
| V _{SUP} | Power Supply Range for VDDD, VDDA | (Note 2) | 2.25 | | 3.63 | V |
| I _{SUP(OFF)} | Supply Current when Powered Down | Software disabled or auto power-down | | 0.1 | 0.5 | μΑ |
| I _{SUP(ON)} | Supply Current of IR Sensing | | | 70 | 90 | μΑ |
| V _I ² _C | Supply Voltage Range for I ² C Interface | | 1.7 | | 3.63 | V |
| fosc | Internal Oscillator Frequency | | 675 | 750 | 825 | kHz |
| t _{int} | ADC Integration/Conversion Time | 16-bit ADC data | | 90 | | ms |
| F _I ² C | I ² C Clock Rate Range | | | 1 to 400 | | kHz |
| DATA_IR0 | Count Output When Dark | E = 0 lux | | 1 | 6 | Counts |
| DATA_FS | Full Scale ADC Code | | | | 65535 | Counts |
| DATA_IR1 | Infrared Count Output | E = 210 lux, Sunlight (Note 3), IR sensing, Range 1 | 15000 | 20000 | 25000 | Counts |
| DATA_IR2 | Infrared Count Output | E = 210 lux, Sunlight (Note 3), IR sensing, Range 2 | | 5000 | | Counts |
| DATA_IR3 | Infrared Count Output | E = 210 lux, Sunlight (Note 3), IR sensing, Range 3 | | 1250 | | Counts |
| DATA_IR4 | Infrared Count Output | E = 210 lux, Sunlight (Note 3), IR sensing, Range 4 | | 312 | | Counts |
| V _{REF} | Voltage of R _{EXT} Pin | | | 0.52 | | V |
| V _{IL} | SCL and SDA Input Low Voltage | | | | 0.55 | V |
| V _{IH} | SCL and SDA Input High Voltage | | 1.25 | | | V |
| I _{SDA,} I _{INT} | SDA and INT Current Sinking Capability | | 4 | 5 | | mA |
| I _{IRDR1} | IRDR Source Current | IS<1:0> = 0 (Note 4) | | 100 | | mA |
| I _{IRDR2} | IRDR Source Current | IS<1:0> = 1 (Note 4) | 44 | 50 | 58 | mA |
| I _{IRDR3} | IRDR Source Current | 15Ω at IRDR pin 15Ω 150 at IRDR pin | | 25 | | mA |
| I _{IRDR4} | IRDR Source Current | IS<1:0> = 3 (Note 4) | | 12.5 | | mA |
| V _{IRLED} | Voltage Head Room of IRDR Pin | | | V _{DD} - 0.6 | | V |
| tr | Rise Time for IRDR Source Current | R_{LOAD} = 15 Ω at IRDR pin, 20% to 80% | | 35 | | ns |
| tf | Fall Time for IRDR Source Current | R_{LOAD} = 15 Ω at IRDR pin, 80% to 20% | | 10 | | ns |
| f _{IRLED1} | IR LED Modulation Frequency | Freq = 0 (Note 4) | | DC | | kHz |
| f _{IRLED2} | IR LED Modulation Frequency | Freq = 1 (Note 4) | | 360 | | kHz |

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Electrical Specifications

 $V_{SUP(VDDD,VDDA)}$ = 3V, T_A = +25°C, R_{EXT} = 499k Ω 1% tolerance, 16-bit ADC operation, unless otherwise specified. (Continued)

| PARAMETER | DESCRIPTION | CONDITION | MIN | TYP | MAX | UNIT |
|---------------------------|-------------------------------------|---|-----|-----|-----|------|
| I _{SUP} (IRLED1) | Supply Current of Proximity Sensing | IS<1:0> = 0, Freq = 0 (Note 4) | | 101 | | mA |
| I _{SUP} (IRLED2) | Supply Current of Proximity Sensing | IS<1:0> = 0, Freq = 1 (Note 4) | | 51 | | mA |
| Duty Cycle | Duty Cycle of IR LED Modulation | | | 50 | | % |
| PROX | | IR and proximity sensing with Range 2; 15Ω @ IRDR pin, IS<1:0> = 0, Freq = 0; E = 210 lux, Sunlight. | | 1.0 | | % |

NOTES:

- V_{SUP} is the common voltage to V_{DDD} and V_{DDA}.
- 3. 850nm infrared LED is used in production test. The 850nm LED irradiance is calibrated to produce the same DATA_IR count against an illuminance level of 210 lux sunlight at sea level.
- 4. See "Register Set" on page 6.

Principles of Operation

Photodiodes and ADC

The ISL29021 contains a photodiode array which converts infrared energy into current. The spectral response for IR sensing is shown in Figure 6 in the performance curves section. After IR radiation is converted to current during the infrared signal processing, the current output is converted to digital by a built-in 16-bit Analog-to-Digital Converter (ADC). An I²C command reads the infrared light intensity in counts.

The converter is a charge-balancing integration type 16-bit ADC. The chosen method for conversion is best for converting small current signals in the presence of an AC periodic noise. A 100ms integration time, for instance, highly rejects 50Hz and 60Hz power line noise simultaneously. See "Integration and Conversion Time" on page 7.

The built-in ADC offers user flexibility in integration time or conversion time. Integration time is determined by an internal oscillator (f_{OSC}), and the n-bit (n = 4, 8, 12,16) counter inside the ADC. A good balancing act of integration time and resolution depending on the application is required for optimal results.

The ADC has I²C programmable range select to dynamically accommodate various IR conditions. For very dim conditions, the ADC can be configured at its lowest range (Range 1). For very bright conditions, the ADC can be configured at its highest range (Range 4) in the proximity sensing.

Low-Power Operation

The ISL29021 initial operation is at the power-down mode after a supply voltage is provided. The data registers contain the default value of 0. When the ISL29021 receives an I²C command to do a one-time measurement from an I²C master, it will start ADC conversion with proximity sensing. It will go to the power-down mode automatically after one conversion is finished and keep the conversion data available for the master to fetch anytime afterwards. The ISL29021 will continuously

do ADC conversion with proximity sensing if it receives an I²C command of continuous measurement. It will continuously update the data registers with the latest conversion data. It will go to the power-down mode after it receives the I²C command of power-down.

Infrared and Proximity Sensing

There are four operational modes in ISL29021: programmable IR sensing once with auto power-down, programmable proximity sensing once with auto power-down, programmable continuous IR sensing and programmable continuous proximity sensing. These four modes can be programmed in series to fulfill the application needs. The detailed program configuration is listed in "Register Set" on page 6.

When the part is programmed for infrared (IR) sensing, the IR light with wavelength within the "IR or Proximity Sensing" spectral response curve on Figure 6 is converted into current. With ADC, the current is converted to an unsigned n-bit (up to 16 bits) digital output.

When the part is programmed for proximity sensing, the external IR LED is turned on by the built-in IR LED driver through the IRDR pin. The amplitude of the IR LED current and the IR LED modulation frequency can be programmed through Command Register II. When the IR from the LED reaches an object and gets reflected back, the reflected IR light with wavelength within the "IR or Proximity Sensing" spectral response curve in Figure 6 is converted into current. With ADC, the current is converted to an unsigned n-bit (up to 16 bits) digital output. The output reading is inversely proportional to the square of the distance between the sensor and the object.

Interrupt Function

The active low interrupt pin is an open drain pull-down configuration. There is also an interrupt bit in the I²C register. The interrupt serves as an alarm or monitoring function to determine whether the infrared light level or the proximity detection level exceeds the upper threshold or goes below the lower threshold. The user can also configure the persistency

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of the interrupt. An unexpected camera flash, for example, can be ignored by setting the persistency to 8 integration cycles.

¹²C Interface

There are eight 8-bit registers available inside the ISL29021. The two command registers define the operation of the device. The command registers do not change until the registers are overwritten. The two 8-bit data Read Only registers are for the ADC output and the Timer output. The data registers contain the ADC's latest digital output, or the number of clock cycles in the previous integration period. The four 8-bit interrupt registers hold 16-bit interrupt high and low thresholds.

The ISL29021's I 2 C interface slave address is internally hardwired as 1000100. When 1000100x with x as R or \overline{W} is sent after the Start condition, this device compares the first seven bits of this byte to its address and matches.

Figure 1 shows a sample one-byte read. Figure 2 shows a sample one-byte write. The I²C bus master always drives the SCL (clock) line, while either the master or the slave can drive the SDA (data) line. Figure 2 shows a sample write. Every I²C transaction begins with the master asserting a start condition (SDA falling while SCL remains high). The following byte is driven by the master, and includes the slave address and read/write bit. The receiving device is responsible for pulling SDA low during the acknowledgement period. Every I²C transaction ends with the master asserting a stop condition (SDA rising while SCL remains high).

For more information about the I^2C standard, please consult the Philips I^2C specification documents.

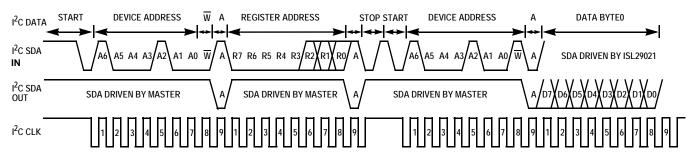


FIGURE 1. I²C READ TIMING DIAGRAM SAMPLE

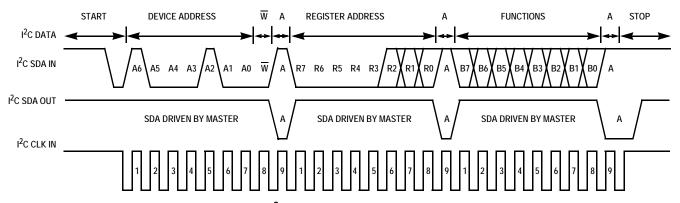


FIGURE 2. I²C WRITE TIMING DIAGRAM SAMPLE

Register Set

There are eight registers that are available in the ISL29021. Table 1 summarizes their functions.

TABLE 1. REGISTER SET

| | | | BIT | | | | | | | |
|------|---------------------|------|------|------|------|------|------|--------|--------|---------|
| ADDR | REG NAME | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | DEFAULT |
| 00h | COMMANDI | OP2 | OP1 | OP0 | 0 | 0 | FLAG | PRST1 | PRST0 | 00h |
| 01h | COMMANDII | 1 | FREQ | IS1 | IS0 | RES1 | RES0 | RANGE1 | RANGE0 | 00h |
| 02h | DATA _{LSB} | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | 00h |
| 03h | DATA _{MSB} | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 | 00h |
| 04h | INT_LT_LSB | TL7 | TL6 | TL5 | TL4 | TL3 | TL2 | TL1 | TL0 | 00h |
| 05h | INT_LT_MSB | TL15 | TL14 | TL13 | TL12 | TL11 | TL10 | TL9 | TL8 | 00h |
| 06h | INT_HT_LSB | TH7 | TH6 | TH5 | TH4 | TH3 | TH2 | TH1 | TH0 | FFh |
| 07h | INT_HT_MSB | TH15 | TH14 | TH13 | TH12 | TH11 | TH10 | TH9 | TH8 | FFh |

Command Register I 00(hex)

The first command register has the following functions:

1. Operation Mode: Bits 7, 6, and 5. These three bits determines the operation mode of the device.

TABLE 2. OPERATION MODE

| BITS 7 TO 5 | OPERATION | |
|-------------|-----------------------|--|
| 000 | Power-down the device | |
| 001 | Reserved (Do not use) | |
| 010 | conce | |
| 011 | roximity once | |
| 100 | Reserved (Do not use) | |
| 101 | Reserved (Do not use) | |
| 110 | IR continuous | |
| 111 | Proximity continuous | |

Interrupt flag; Bit 2. This is the status bit of the interrupt.
 The bit is set to logic high when the interrupt thresholds have been triggered, and logic low when not yet triggered.
 Once triggered, INT pin stays low and the status bit stays high. Both interrupt pin and the status bit are automatically cleared at the end of Command Register I transfer.

TABLE 3. INTERRUPT FLAG

| BIT 2 OPERATION | | |
|--------------------------|---|--|
| 0 | 0 Interrupt is cleared or not triggered yet | |
| 1 Interrupt is triggered | | |

 Interrupt persist; Bits 1 and 0. The interrupt pin and the interrupt flag is triggered/set when the data sensor reading is out of the interrupt threshold window after m consecutive number of integration cycles. The interrupt persist bits determine m.

TABLE 4. INTERRUPT PERSIST

| BITS 1 TO 0 | NUMBER OF INTEGRATION CYCLES |
|-------------|------------------------------|
| 00 | 1 |
| 01 | 4 |
| 10 | 8 |
| 11 | 16 |

Command Register II 01(hex)

The second command register has the following functions:

Proximity Sensing Scheme: Bit 7. This bit programs the function of the proximity detection. Logic 1 of this bit, Scheme 1, makes n-1 (3, 7, 11, 15) bits (2's complementary) proximity_less_ambient detection. The range of Scheme 1 proximity count is from -2⁽ⁿ⁻¹⁾ to 2⁽ⁿ⁻¹⁾, Scheme 1 proximity detection is less affected by the ambient IR noise variation.

TABLE 5. PROXIMITY SENSING SCHEME

| BIT 7 | OPERATION | | | |
|-------|---|--|--|--|
| 0 | Reserved | | | |
| 1 | Sensing IR from LED with ambient IR rejection | | | |

Modulation Frequency: Bits 6. This bit sets the IR LED driver's modulation frequency.

TABLE 6. MODULATION FREQUENCY

| BITS 6 | MODULATION FREQUENCY (kHz) |
|--------|-------------------------------|
| 0 | DC |
| 1 | 360 |

 Amplitude of IR driver current: Bits 5 and 4. This device provides current source to drive an external IR LED. The drive capability can be programmed through Bits 5 and 4. For example, the device sources 12.5mA out of the IRDR pin if Bits 5 and 4 are 0.

TABLE 7. CURRENT SOURCE CAPABILITY AT IRDR PIN

| BITS 5 TO 4 | IRDR PIN SOURCE CURRENT |
|-------------|-------------------------|
| 00 | 12.5mA IR LED driver |
| 01 | 25mA IR LED driver |
| 10 | 50mA IR LED driver |
| 11 | 100mA IR LED driver |

4. Resolution: Bits 3 and 2. Bits 3 and 2 determine the ADC's resolution and the number of clock cycles per conversion in Internal Timing Mode. Changing the number of clock cycles does more than just change the resolution of the device. It also changes the integration time, which is the period the device's analog-to-digital (A/D) converter samples the photodiode current signal for a measurement.

TABLE 8. RESOLUTION/WIDTH

| BITS 3 TO 2 | NUMBER OF CLOCK CYCLES | n-BIT ADC |
|-------------|------------------------|-----------|
| 00 | 2^{16} = 65,536 | 16 |
| 01 | $2^{12} = 4,096$ | 12 |
| 10 | 2 ⁸ = 256 | 8 |
| 11 | 2 ⁴ = 16 | 4 |

5. Range: Bits 1 and 0. The Full Scale Range (FSR) can be adjusted via I^2C using Bits 1 and 0. Table 9 lists the possible values of FSR for the $499k\Omega$ R_{FXT} resistor.

TABLE 9. RANGE/FSR

| BITS 1:0 | k | RANGE(k) | FSR @ IR SENSING |
|----------|---|----------|------------------|
| 00 | 1 | Range1 | Refer to page 3 |
| 01 | 2 | Range2 | Refer to page 3 |
| 10 | 3 | Range3 | Refer to page 3 |
| 11 | 4 | Range4 | Refer to page 3 |

Data Registers (02 hex and 03 hex)

The device has two 8-bit read-only registers to hold the data from LSB to MSB for ADC. The most significant bit (MSB) is accessed at 03 hex, and the least significant bit (LSB) is accessed at 02 hex. For 16-bit resolution, the data is from D0 to D15; for 12-bit resolution, the data is from D0 to D11; for 8-bit resolution, the data is from D0 to D7. The registers are refreshed after every conversion cycle.

TABLE 10. DATA REGISTERS

| ADDRESS (hex) | CONTENTS |
|------------------|---|
| 02 | D0 is LSB for 4, 8, 12 or 16-bit resolution, D3 is MSB for 4-bit resolution, D7 is MSB for 8-bit resolution |
| 03 | D15 is MSB for 16-bit resolution, D11 is MSB for 12-bit resolution |

Interrupt Registers (04, 05, 06 and 07 hex)

Registers 04 and 05 hex set the low (LO) threshold for the interrupt pin and the interrupt flag. 04 hex is the LSB and 05 hex is the MSB. By default, the Interrupt threshold LO is 00 hex for both LSB and MSB.

Registers 06 and 07 hex set the high (HI) threshold for the interrupt pin and the interrupt flag. 06 hex is the LSB and 07 hex is the MSB. By default, the Interrupt threshold HI is FF hex for both LSB and MSB.

Integration and Conversion Time

The ADC resolution and $f_{\mbox{OSC}}$ determines the integration time, $t_{\mbox{int}}$.

$$t_{int} = 2^{n} \times \frac{1}{t_{OSC}} = 2^{n} \times \frac{R_{EXT}}{725kHz \times 499k\Omega}$$
 (EQ. 1)

where n is the number of bits of resolution and n = 4, 8, 12 or 16. 2^n , therefore, is the number of clock cycles. n can be programmed at the command register 01(hex) bits 3 and 2.

TABLE 11. INTEGRATION TIME OF n-BIT ADC

| $\mathbf{R}_{\mathbf{EXT}}$ ($\mathbf{k}\Omega$) | n = 16-BIT | n = 12-BIT | n = 8-BIT | n = 4-BIT |
|--|------------|------------|-----------|-----------|
| 250 | 45ms | 2.812ms | 175.5µs | 10.8µs |
| 499** | 90ms | 5.63ms | 351µs | 21.6µs |

^{**}Recommended R_{FXT} resistor value

External Scaling Resistor R_{EXT} for f_{OSC} and Range

The ISL29021 uses an external resistor R_{EXT} to fix its internal oscillator frequency, f_{OSC} . Range. f_{OSC} and Range are inversely proportional to R_{EXT} . For user simplicity, the proportionality constant is referenced to 499k Ω :

$$Range = \frac{499k\Omega}{R_{FXT}} \times Range(k)$$
 (EQ. 2)

$$f_{OSC} = \frac{499 k\Omega}{R_{EXT}} \times 725 kHz$$
 (EQ. 3)

Noise Rejection

In general, integrating type ADC's have excellent noise-rejection characteristics for periodic noise sources whose frequency is an integer multiple of the conversion rate. For instance, a 60Hz AC unwanted signal's sum from 0ms to k*16.66ms (k = 1,2...k_i) is zero. Similarly, setting the device's integration time to be an integer multiple of the periodic noise signal, greatly improves the proximity sensor output signal in the presence of noise.

ADC Output in IR Sensing

The ISL29021's ADC output codes, DATA, are directly proportional to the IR intensity received in the IR sensing.

$$DATA_{IR} = \beta \times E_{IR}$$
 (EQ. 4)

Here, E_{IR} is the received IR intensity. The constant β changes with the spectrum of background IR noise like sunlight and incandescent light. The β also changes with the ADC's range and resolution selections.

ADC Output in Proximity Sensing

In the proximity sensing, the ADC output codes, DATA, are directly proportional to the total IR intensity from the

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background IR noise and from the IR LED driven by the ISL29021 as shown in Equation 5.

$$DATA_{PROX} = \beta \times E_{IR} + \gamma \times E_{IFD}$$
 (EQ. 5)

Here, β and E_{IR} have the same meanings as in Equation 4. The constant γ depends on the spectrum of the used IR LED and the ADC's range and resolution selections. E_{LED} is the IR intensity which is emitted from the IR LED and reflected by a specific objector to the ISL29021. E_{LED} depends on the current to the IR LED and the surface of the object. E_{LED} decreases with the square of the distance between the object and the sensor.

If background IR noise is small, E_{IR} can be neglected, and the ADC output directly decreases with the distance. If there is significant background IR noise, ISL29021 is to do a proximity sensing using Scheme 1 to do on-chip background IR noise subtraction.

Figure 9 shows ISL29021 configured at 12-bit ADC resolution and sensitivity range select at 16000 (range 3) for the proximity reading. A 12.5mA external LED current at 360kHz modulation frequency detects three different sensing objects: 92% brightness paper, 18% gray card and ESD black foam. Figure 10 shows ISL29021 configured at 12-bit ADC resolution and sensitivity range select at 1000 (range 1) for the proximity reading, with a programmed external LED at 360kHz modulation frequency, detecting the same sensing object: 18% gray card under four different external LED current: 12.5mA, 25mA, 50mA and 100mA to compare the proximity readout versus distance.

ISL29021 Proximity sensing relies on the amount of IR reflected back from the objects to be detected. Clearly, it can not detect an optically black object that reflects no light. However, ISL29021 is sensitive enough to detect a black ESD foam, which reflects slightly less than 1% of IR, as shown in Figure 9. For biological objects, blonde hair reflects more than brunette hair, as expected and shown in Figure 11. Also notice that skin tissue is much more reflective than hair. IR penetrates into the skin and is reflected or scattered back from within. As a result, the proximity count peaks at contact and monotonically decreases as skin moves away. This characteristic is very different from that of a plain paper reflector.

Interrupt Function

Depending on the mode of operation set by Bits 7, 6 and 5 of command register 00 hex, the upper and lower interrupt thresholds are for either infrared signal level or proximity detection. After each change of mode of operation, it is expected a new set of thresholds are loaded to interrupt registers 04, 05, 06 and 07 hex for proper interrupt detection. Also, the interrupt persist counter will be reset to 0 when the mode of operation is changed.

LED Modulation for Proximity Detection

ISL29021 offers two ways to modulate the LED in the Proximity Detection mode - DC or 360kHz (with 50% duty cycle) by bit 6 of register 01h. At the IRDR pin, there are four different IRDR LED currents; 12.5, 25, 50, and 100mA outputs selectable by bits 4 and 5 of register 01h. With the LED running in the DC mode, the proximity detection is twice as sensitive but consumes 2x more current. The sensitivity of LED 50mA, DC 50mA is identical to that of 100mA, 360kHz modulation. Please note that the ISL29021 does not include a LED.

Current Consumption Estimation

The low power operation is achieved through sequential readout in the serial fashion, as shown in Figure 3, the device requires three different phases in serial during the entire detection cycle to do infrared sensing and proximity sensing. The external IR LED will only be turned on during the proximity sensing phase under user program controlled current at modulated frequency depends on user selections. Figure 3 also shows the current consumption during each IR sensing and Proximity sensing phase. For example, at 8-bit ADC resolution the integration time is 0.4ms. If user programed 50mA current to supply external IR LED at 360kHz modulated frequency, during the entire operation cycle that includes IR sensing and Proximity sensing three different serial phases, the detection occurs once every 30ms, the average current consumption including external IR LED drive current can be calculated from Equation 6:

$$[(0.05\text{mA} + 0.05\text{mA} + 1\text{mA} + (50\text{mA}*50\%))*0.4\text{ms})]/30\text{ms} = 0.35\text{mA}$$
 (EQ. 6)

If at a 12-bit ADC resolution where the integration time for each serial phase becomes 7ms and the total detection time becomes 100ms, the average current can be calculated from Equation 7:

$$[(0.05\text{mA} + 0.05\text{mA} + 1\text{mA} + (50\text{mA}*50\%))*7\text{ms})]/100\text{ms} = 1.83\text{mA}$$
 (EQ. 7)

Suggested PCB Footprint

It is important that the users check the "Surface Mount Assembly Guidelines for Optical Dual FlatPack No Lead (ODFN) Package" before starting ODFN product board mounting.

http://www.intersil.com/data/tb/TB477.pdf

Layout Considerations

The ISL29021 is relatively insensitive to layout. Like other I²C devices, it is intended to provide excellent performance even in significantly noisy environments. There are only a few considerations that will ensure best performance.

Route the supply and I^2C traces as far as possible from all sources of noise. Use two power-supply decoupling capacitors $1\mu F$ and $0.1\mu F$, placed close to the device.

intersil® FN6732.0 March 3. 2009

Typical Circuit

A typical application for the ISL29021 is shown in Figure 4. The ISL29021's I^2C address is internally hardwired as 1000100. The device can be tied onto a system's I^2C bus together with other I^2C compliant devices.

Soldering Considerations

Convection heating is recommended for reflow soldering; direct-infrared heating is not recommended. The plastic ODFN package does not require a custom reflow soldering profile, and is qualified to +260°C. A standard reflow soldering profile with a +260°C maximum is recommended.

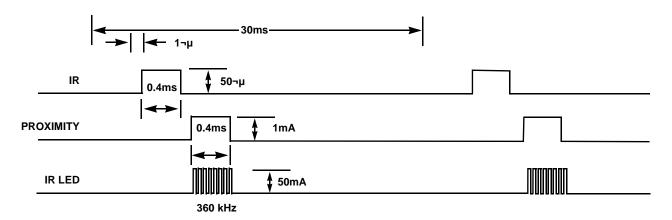


FIGURE 3. CURRENT CONSUMPTION FOR EACH INTEGRATION PHASE AND DETECTION CYCLE

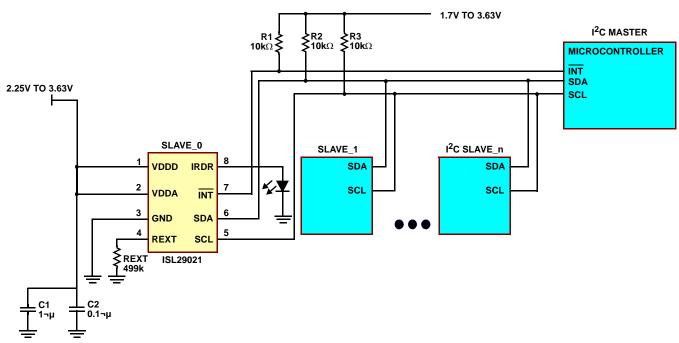


FIGURE 4. ISL29021 TYPICAL CIRCUIT

Typical Performance Curves $V_{SUP}(V_{DDD}, V_{DDA}) = 3V$, $R_{EXT} = 499k\Omega$

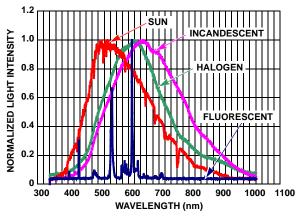


FIGURE 5. SPECTRUM OF FOUR LIGHT SOURCES

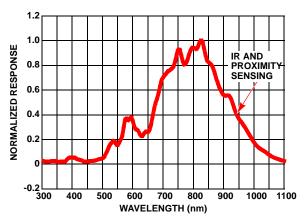


FIGURE 6. SPECTRAL RESPONSE FOR PROXIMITY SENSING

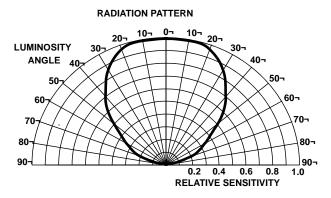


FIGURE 7. RADIATION PATTERN

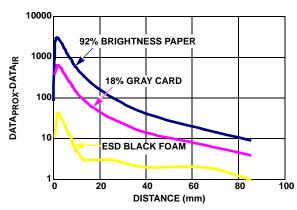


FIGURE 8. ADC OUTPUT vs DISTANCE WITH DIFFERENT OBJECTS IN PROXIMITY SENSING

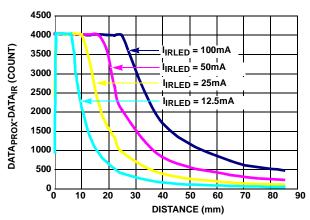


FIGURE 9. ADC OUTPUT vs DISTANCE WITH DIFFERENT LED CURRENT AMPLITUDES IN PROXIMITY SENSING

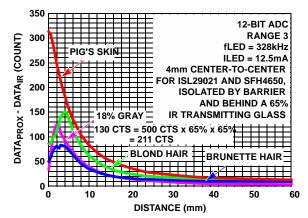


FIGURE 10. PROXIMITY DETECTIONS OF VARIOUS BIOLOGICAL OBJECTS

Typical Performance Curves $V_{SUP}(V_{DDD}, V_{DDA}) = 3V$, $R_{EXT} = 499k\Omega$ (Continued)

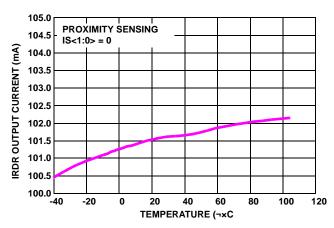


FIGURE 11. OUTPUT CURRENT vs TEMPERATURE IN PROXIMITY SENSING

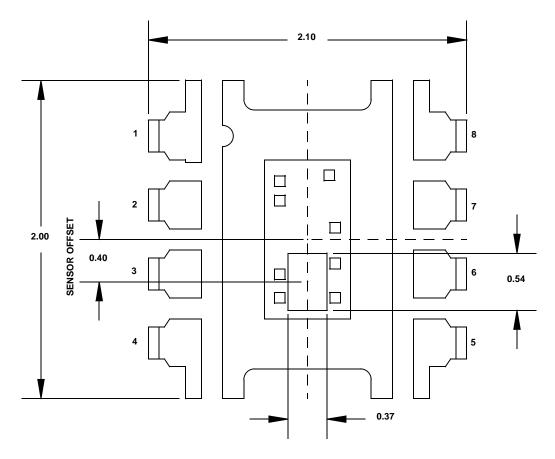


FIGURE 12. 8 LD ODFN SENSOR LOCATION OUTLINE

For additional products, see www.intersil.com/en/products.html

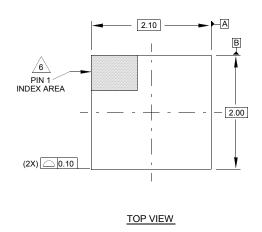
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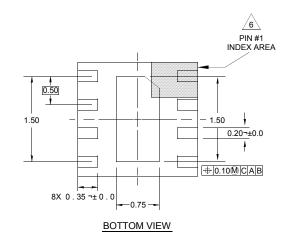
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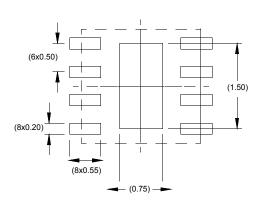
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Package Outline Drawing

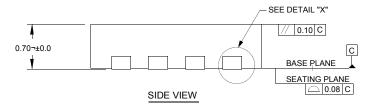
L8.2.1x2.0 8 LEAD OPTICAL DUAL FLAT NO-LEAD PLASTIC PACKAGE (ODFN) Rev 0, 10/08

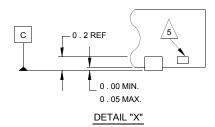






TYPICAL RECOMMENDED LAND PATTERN





NOTES:

- Dimensions are in millimeters.
 Dimensions in () for Reference Only.
- 2. Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
- 3. Unless otherwise specified, tolerance : Decimal $\neg \pm 0.0$
- 4. Dimension b applies to the metallized terminal and is measured between 0.25mm and 0.35mm from the terminal tip.
- 5. Tiebar shown (if present) is a non-functional feature.
- The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 indentifier may be either a mold or mark feature.

